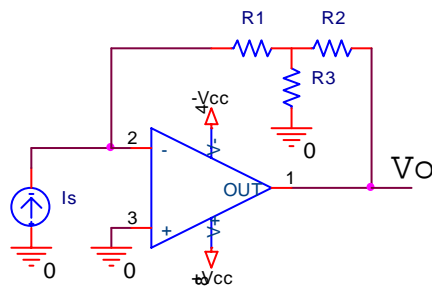


SENSORES OPTICOS (enero-2005)

Para acondicionar sensores que proporcionan salida en corriente con una pequeña sensibilidad es necesario utilizar resistencias de valor elevado para obtener la suficiente ganancia. La “red en T” permite obtener resistencias equivalentes de valor elevado sin necesidad de utilizar componentes de gran valor óhmico.



- Determinar la expresión de la resistencia equivalente de la red en T.
- Determina el error absoluto a la salida asociado a la tensión offset del A.O.
- Determina el error absoluto a la salida asociado a las corrientes de polarización de entrada del A.O.

SENSORES OPTICOS

Se dispone de un fotodiodo que tiene una sensibilidad de $0.43 \mu\text{A}/\mu\text{W}$, una capacidad parásita de 20 pF y una resistencia de fugas de $100 \text{ M}\Omega$. Se conecta a un amplificador de transimpedancia con $R = 10 \text{ M}\Omega$ y $C = 2 \text{ pF}$, basado en el TLC2201B, alimentado entre 0 y 5 V . Si a la salida se conecta un filtro que deja pasar la banda entre 10 Hz y 10 kHz , ¿Cuál es la tensión de ruido a la salida?

TLC220x, TLC220xA, TLC220xB, TLC220xY
Advanced LinCMOS™ LOW-NOISE PRECISION
OPERATIONAL AMPLIFIERS

SLOS175 – FEBRUARY 1997

TLC220x, TLC220xA, TLC220xB, TLC220xY
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OPERATIONAL AMPLIFIERS

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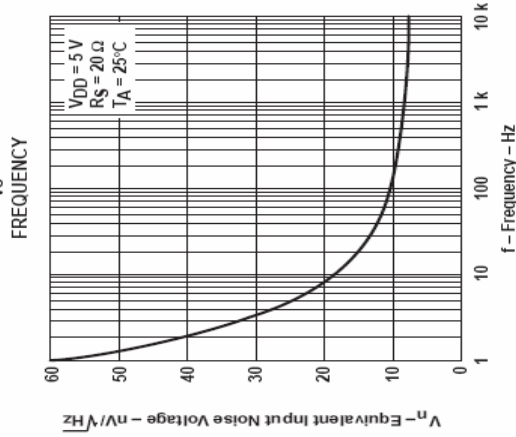
- **B Grade Is 100% Tested for Noise**
30 nV/√Hz Max at f = 10 Hz
12 nV/√Hz Max at f = 1 kHz
- **Low Input Offset Voltage . . . 500 μV Max**
- **Excellent Offset Voltage Stability**
With Temperature . . . 0.5 μV/°C Typ
- **Rail-to-Rail Output Swing**
- **Low Input Bias Current**
1 pA Typ at TA = 25°C
- **Common-Mode Input Voltage Range**
Includes the Negative Rail
- **Fully Specified For Both Single-Supply and Split-Supply Operation**

description

The TLC220x, TLC220xA, TLC220xB, and TLC220xY are precision, low-noise operational amplifiers using Texas Instruments Advanced LinCMOS™ process. These devices combine the noise performance of the lowest-noise JFET amplifiers with the dc precision available previously only in bipolar amplifiers. The Advanced LinCMOS™ process uses silicon-gate technology to obtain input offset voltage stability with temperature and time that far exceeds that obtainable using metal-gate technology. In addition, this technology makes possible input impedance levels that meet or exceed levels offered by top-gate JFET and expensive dielectric-isolated devices.

The combination of excellent dc and noise performance with a common-mode input voltage range that includes the negative rail makes these devices an ideal choice for high-impedance, low-level signal-conditioning applications in either single-supply or split-supply configurations.

TYPICAL EQUIVALENT
 INPUT NOISE VOLTAGE
 vs
 FREQUENCY



TLC22011 electrical characteristics at specified free-air temperature, VDD = 5 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TA †	TLC2201AI			TLC2201BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	25°C	80	200	80	200	200	μA	
αV _{IO}	Temperature coefficient of input offset voltage	Full range	350			350			μV/°C
		Full range	0.5			0.5			
I _{IO}	Input offset voltage long-term drift (see Note 4)	25°C	0.001			0.005			μV/mo
I _{IO}	Input offset current	25°C	0.5			0.5			pA
		Full range	1			1			
I _{IB}	Input bias current	25°C	1			1			pA
		Full range	150			150			
V _{ICR}	Common-mode input voltage range	Full range	0 to 2.7			0 to 2.7			V
V _{OH}	Maximum high-level output voltage	25°C	4.7			4.7			V
V _{OL}	Maximum low-level output voltage	25°C	0			0			mV
		Full range	50			50			
A _V D	Large-signal differential voltage amplification	25°C	150			150			V/mV
		Full range	100			100			
		25°C	25			25			
		Full range	15			15			
CMRR	Common-mode rejection ratio	25°C	90			90			dB
KSVR	Supply voltage rejection ratio (ΔV _{DD} ± ΔV _{IO})	25°C	85			85			dB
		Full range	90			90			
I _{DD}	Supply current	25°C	1			1			mA
		Full range	1.5			1.5			

TLC22011 operating characteristics at specified free-air temperature, VDD = 5 V

PARAMETER	TEST CONDITIONS	TA †	TLC2201AI			TLC2201BI			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
SR	Stew rate at unity gain	25°C	1.8			1.8			V/μs
		Full range	1.2			1.2			
V _n	Equivalent input noise voltage (see Note 5)	25°C	18			18			nV/√Hz
		25°C	8			8			
V _{N(PP)}	Peak-to-peak equivalent input noise voltage	25°C	0.5			0.5			μV
		25°C	0.7			0.7			
I _n	Equivalent input noise current	25°C	0.6			0.6			fA/√Hz
		25°C	1.8			1.8			
φ _m	Phase margin at unity gain	25°C	45°			45°			°
		25°C	45°			45°			

† Full range is -40°C to 85°C.

NOTE 5: This parameter is tested on a sample basis for the TLC2201A and on all devices for the TLC2201B. For other test requirements, please contact the factory. This statement has no bearing on testing or non-testing of other parameters.

SENSORES OPTICOS

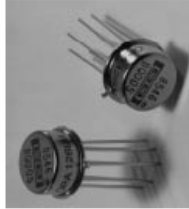
Se desea acondicionar la señal del fotodiodo HP 5082-4204 mediante un amplificador corriente-tensión (transimpedancia -R con una pequeña C en paralelo en la realimentación-) usando el A.O. OPA128J de Burr-Brown, cuyas características se adjuntan.

A 770 nm, la sensibilidad del fotodiodo es de $0.5 \mu\text{A}/\mu\text{W}$, su resistencia de fugas es de $100 \text{ G}\Omega$, y su capacidad cuando no se polariza inversamente es de 6.5 pF . Entre el A.O. y el montaje se añaden 3.5 pF más.

Si el A.O. se supone ideal, ¿cuánto debe valer R para tener a la salida una sensibilidad de $1 \text{ V}/\mu\text{W}$?

¿Qué error se produce si, ignorando que la ganancia en lazo abierto del A.O es limitada, suponemos que a 10 kHz la sensibilidad es la misma que a baja frecuencia?

¿Qué error produce a 10 kHz la limitada resistencia de entrada del A.O?



Difet® Electrometer-Grade OPERATIONAL AMPLIFIER

FEATURES

- ULTRA-LOW BIAS CURRENT: 75fA max
- LOW OFFSET: 500µV max
- LOW DRIFT: 5µV/°C max
- HIGH OPEN-LOOP GAIN: 110dB min
- HIGH COMMON-MODE REJECTION: 90dB min
- IMPROVED REPLACEMENT FOR AD515 AND AD549

APPLICATIONS

- ELECTROMETER
- MASS SPECTROMETER
- CHROMATOGRAPH
- ION GAUGE
- PHOTODETECTOR
- RADIATION-HARD EQUIPMENT

DESCRIPTION

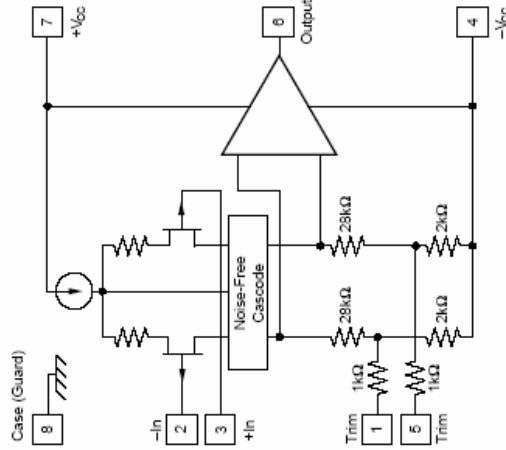
The OPA128 is an ultra-low bias current monolithic operational amplifier. Using advanced geometry dielectrically-isolated FET (*Difet*®) inputs, this monolithic amplifier achieves a performance level exceeding even the best hybrid electrometer amplifiers.

Laser-trimmed thin-film resistors give outstanding voltage offset and drift performance.

A noise-free cascode and low-noise processing give the OPA128 excellent low-level signal handling capabilities. Flicker noise is very low.

The OPA128 is an improved pin-for-pin replacement for the AD515.

Difet® Burr-Brown Corp.



OPA128 Simplified Circuit

SPECIFICATIONS ELECTRICAL

At $V_{CC} = \pm 15\text{VDC}$ and $T_A = +25^\circ\text{C}$ unless otherwise noted. Pin 8 connected to ground.

PARAMETER	CONDITIONS	OPA128JUM			OPA128KLM			OPA128SM			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
BIAS CURRENT⁽¹⁾ Input Bias Current	$V_{CM} = 0\text{VDC}$, $R_L \geq 10\text{k}\Omega$	± 150	± 300	± 150	± 75	± 150	± 75	± 150	± 75	± 150	fA
OFFSET CURRENT⁽¹⁾ Input Offset Current	$V_{CM} = 0\text{VDC}$, $R_L \geq 10\text{k}\Omega$	65			30			30			fA
OFFSET VOLTAGE⁽¹⁾ Input Offset Voltage Average Drift Supply Rejection	$V_{CM} = 0\text{VDC}$ $T_A = T_{MIN}$ to T_{MAX}	± 500 ± 20 80	± 1000 ± 20 ± 1	± 500 ± 10 ± 32	± 140 ± 5 ± 1	± 140 ± 5 ± 1	± 140 ± 5 ± 1	± 140 ± 5 ± 1	± 140 ± 5 ± 1	± 500 ± 10 ± 32	μV $\mu\text{V}/^\circ\text{C}$ $\mu\text{V}/\text{V}$
NOISE Voltage: $f_o = 10\text{Hz}$ $f_o = 100\text{Hz}$ $f_o = 1\text{kHz}$ $f_o = 10\text{kHz}$ $f_o = 0.1\text{Hz}$ to 10Hz Current: $f_o = 0.1\text{Hz}$ to 10Hz $f_o = 0.1\text{Hz}$ to 20kHz		92 78 27 15 2.4 4 4.2 0.22		92 78 27 15 2.4 4 3 0.16		92 78 27 15 2.4 4 3 0.16		92 78 27 15 2.4 4 3 0.16			nV/√Hz nV/√Hz nV/√Hz µV/rms µV/p-p fA/p-p fA/√Hz
IMPEDANCE Differential Common-Mode		$10^{13} \parallel 1$ $10^{15} \parallel 2$		$10^{13} \parallel 1$ $10^{15} \parallel 2$		$10^{13} \parallel 1$ $10^{15} \parallel 2$		$10^{13} \parallel 1$ $10^{15} \parallel 2$		$10^{13} \parallel 1$ $10^{15} \parallel 2$	$\Omega \parallel \text{pF}$ $\Omega \parallel \text{pF}$
VOLTAGE RANGE⁽⁴⁾ Common-Mode Input Range Common-Mode Rejection	$V_{IN} = \pm 10\text{VDC}$	± 10 80	± 12 118	± 10 80	± 10 90	± 12 118	± 10 90	± 12 118	± 10 90	± 12 118	V dB
Open-Loop Voltage Gain	$R_L \geq 2\text{k}\Omega$	94	128	110	128	110	128	110	128	128	dB
Unity Gain, Small Signal Full Power Response	20V/p-p , $R_L = 2\text{k}\Omega$	0.5	1	0.5	1	0.5	1	0.5	1	0.5	MHz
Slew Rate	$V_O = \pm 10\text{V}$, $R_L = 2\text{k}\Omega$	0.5	3	1	3	1	3	1	3	3	V/µs
Settling Time, 0.1% 0.01% Overload Recovery, 50% Overdrive⁽³⁾	Gain = -1, $R_L = 2\text{k}\Omega$ 10V Step Gain = -1	10	10	10	10	10	10	10	10	10	µs
Voltage Output	$R_L = 2\text{k}\Omega$	± 10	± 13	± 10	± 10	± 13	± 10	± 13	± 10	± 13	V
Current Output	$V_O = \pm 10\text{VDC}$ DC, Open Loop	± 5	± 10	± 5	± 5	± 10	± 5	± 10	± 5	± 10	mA
Output Resistance	DC, Open Loop	100	1000	100	1000	100	1000	100	1000	1000	Ω
Load Capacitance Stability	Gain = +1	10	34	55	10	34	55	10	34	55	pF
Short Circuit Current		10	34	55	10	34	55	10	34	55	mA
Rated Voltage											VDC
Voltage Range, Derated Performance											VDC
Current, Quiescent	$I_O = \text{Dm-ADC}$	± 5	± 15	± 5	± 5	± 15	± 5	± 15	± 5	± 15	mA
Specification	Ambient Temp. Storage	0	+70	-55	-55	+125	-55	-65	+125	-55	°C
θ Junction-Ambient	Ambient Temp.	0	+70	-55	-65	+125	-55	-65	+125	-55	°C/W

NOTES: (1) Offset voltage, offset current, and bias current are measured with the units fully warmed up. Bias current doubles approximately every 11°C. (2) Sample tested. (3) Overload recovery is defined as the time required for the output to return from saturation to linear operation following the removal of a 50% input overdrive. (4) If it is possible for the input voltage to exceed the supply voltage, a series protection resistor should be added to limit input current to 0.5mA. The input devices can withstand overload currents of 0.3mA indefinitely without damage.